EE 321 Analog Electronics, Fall 2012 Exam 3 November 30, 2012 solution

This is a closed-book exam. Calculators allowed. The exam is designed for conceptual understanding not long derivations. You MUST box your answer. When there are multiples values in an answer summarize them in a single box. Correct answer boxed and derivation gives you 10 points. Either is 5 points. Neither is 0 points.

1. Find i_D , v_D , v_S , and v_G for this circuit, with $V_t = 1 \text{ V}$, and $k'_n \frac{W}{L} = 1 \text{ mA}/\text{V}^2$



2. Replace the $4\,k\Omega$ with a $10\,k\Omega$ and find the same quantities.

GVESS SATURATION MODE. THEN NO IS THE SAME, AND NJ=V00-NDRD = 10-0.76·10=2.4V. NJ IS ALSO THE SATTE, SO WE FIND THAT NJSNES-VI WHICH VERIHES SATURATION MODE.

$$N_{5}=0.76V$$
 $N_{6}=3V$ $N_{5}=2.4V$ $N_{0}=0.76mA$

3. Make a MOSFET current mirror which provides 1 mA to a load connected between ground and the current mirror output. Assume $V_t = 1 \text{ V}$, and $k'_p \frac{W}{L} = 1 \text{ mA/V}^2$, and 10 V supplies.



4. For the current mirror in the previous circuit what is the largest value of the load resistor that will maintain saturation mode operation of the MOSFETs?

MAX VOLTAGE IS
$$10V - (V_{S_{c}} - V_{A}) = 10V - 1.91Y = 8.55V$$

MAX (NESISTANCE
 $P_{LMAX} = 8.59 \text{ B.C}$

5. Pick resistors in this circuit such that a CS amplifier is implemented (not a CS with R_S) which has (a) output biased at 5 V, (b) $A_{vo} = -10$, (c) smallest gate resistor of 1 MΩ. Assume $k'_n \frac{W}{L} = 1 \frac{\text{mA}}{\text{V}^2}$ and $V_t = 1 \text{ V}$.

$$R_{G1} = +10 V \qquad R_{S} = 0 \qquad A_{ND} = -\frac{2I_{D}R_{D}}{V_{6S} - V_{4}} \\ = 2V \\ = R_{C} \qquad I_{D}R_{D} = 5V = > V_{5S} = -\frac{2I_{D}R_{D}}{A_{ND}} + V_{4} = -\frac{10}{-10} + 1 \\ = 2V \\ = >R_{C2} = 1/ML, R_{C1} = 5/ML \\ = 0.5 mA \\ = >R_{D} = 10 \text{ kl} \\ R_{C1} = 5/ML \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = 1/ML \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1 \\ R_{C1} = -\frac{10}{10} + 1 \\ R_{C2} = -\frac{10}{10} + 1$$

6. Draw the small-signal model for this amplifier, adding also a source (with $R_{\rm src}$) and a load (R_L) .



7. If $R_{G1} = 200 \text{ k}\Omega$, $R_{G2} = 800 \text{ k}\Omega$, $R_S = 1 \text{ k}\Omega$, $R_D = 5 \text{ k}\Omega$, $V_t = 1 \text{ V}$, $k'_n \frac{W}{L} = 1 \text{ mA/V}^2$, $R_{\text{src}} = 100 \text{ k}\Omega$, and $R_L = 3 \text{ k}\Omega$, what are A_{vo} , A_v , G_v , R_{in} , and R_{out} ?

$$\begin{split} \mathcal{J}_{M} &= \frac{2I_{D}}{V_{S}-V_{t}} \qquad \begin{array}{l} I_{D} = \frac{1}{2} \left(V_{c} - V_{k} - I_{D} R_{S} \right)^{2} \\ \mathcal{J}_{M} &= \frac{2I_{D}}{V_{S}-V_{t}} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} 2I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} R_{s} \left(V_{s} + V_{k} \right) \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} \\ &= \frac{2 \cdot 0.27}{I} \qquad \end{array} \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} I_{D} = (V_{c} - V_{k})^{2} + I_{D}^{2} R_{s}^{2} - 2I_{D} \\ &= \frac{1}{2} \qquad \end{array} \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} I_{D} = (V_{c} - V_{L})^{2} \\ &= \frac{1}{2} \qquad \end{array} \\ &= \frac{2 \cdot 0.27}{I} \qquad \end{array} \\ &= \frac{2 \cdot 0.27}{I} \qquad \begin{array}{l} R_{Avv} = -\frac{9}{I} R_{b} R_{$$