

Lecture 3

Probability and Stochastic Processes

EE 521: Instrumentation and Measurements

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1 Review Material

1.1 Signal Classification

Assume the voltage across a resistor R is $e(t)$ and is producing a current $i(t)$. The instantaneous power per ohm is $p(t) = e(t)i(t)/R = i^2(t)$.

Total Energy

$$E = \lim_{T \rightarrow \infty} \int_{-T}^T i^2(t) dt \quad (1)$$

Average Power

$$P = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T i^2(t) dt \quad (2)$$

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Arbitrary signal $x(t)$

Total Normalized Energy

$$E \triangleq \lim_{T \rightarrow \infty} \int_{-T}^T |x(t)|^2 dt = \int_{-\infty}^{\infty} |x(t)|^2 dt \quad (3)$$

Normalized Power

$$P \triangleq \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt \quad (4)$$

- $x(t)$ is an *energy signal* iff $0 < E < \infty$, so that $P = 0$.
- $x(t)$ is a *power signal* iff $0 < P < \infty$, so that $E = \infty$.

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1.2 Time Averages

For Energy Signals

$$\phi(\tau) = \int_{-\infty}^{\infty} x(\lambda)x(\lambda + \tau) d\lambda \quad (5)$$

Provides a measure of similarity or coherence between a signal and a delayed version of itself.
Note that $\phi(0) = E$

For Power Signals

$$R(\tau) = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t)x(t + \tau) dt \quad (6)$$

For Periodic Signals

$$R(\tau) = \frac{1}{T_0} \int_{T_0} x(t)x(t + \tau) dt \quad (7)$$

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1.3 Frequency Domain

Fourier Transform Equations

$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt \quad (8)$$

$$x(t) = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df \quad (9)$$

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Energy Spectral Density

Rayleigh's Energy Theorem or Parseval's theorem

$$E = \int_{-\infty}^{\infty} |x(t)|^2 dt = \int_{-\infty}^{\infty} |X(f)|^2 df \quad (10)$$

Energy Spectral Density

$$G(f) \triangleq |X(f)|^2 \quad (11)$$

with units of $\text{volts}^2\text{-sec}^2$ or, if considered on a per-ohm basis, $\text{watts-sec/Hz} = \text{joules/Hz}$

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Power Spectral Density

$$P = \int_{-\infty}^{\infty} S(f)df = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt \quad (12)$$

where we define $S(f)$ as the power spectral density with units of watts/Hz. Note that $R(0) = \int_{-\infty}^{\infty} S(f)df$.

Proof

In an analogy to the energy signals, let us define a function that would give us some indication of the relative power contributions at various frequencies, as $S_x(\omega)$. This function has units of power per Hz and its integral yields the power in $x(t)$ and is known as *power spectral density* function. Mathematically,

$$P = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_x(\omega) d\omega. \quad (13)$$

Assume that we are given a signal $x(t)$ and we truncate it over the interval $(-T/2, T/2)$. This truncated version is $x(t)\Pi(t/T)$. If $x(t)$ is finite over the interval $(-T/2, T/2)$, then the truncated function $x(t)\Pi(t/T)$ has finite energy and its Fourier transform $X_T(\omega)$ is

$$X_T(\omega) = \mathcal{F}\{x(t)\Pi(t/T)\}. \quad (14)$$

Parseval's theorem of the truncated version is

$$\int_{-T/2}^{T/2} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X_T(\omega)|^2 d\omega. \quad (15)$$

Therefore, the average power P across a one-ohm resistor is given by

$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\infty} |X_T(\omega)|^2 d\omega. \quad (16)$$

Combining Equations (1) and (4), we get

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} S_x(\omega) d\omega = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\infty} |X_T(\omega)|^2 d\omega. \quad (17)$$

In addition if we insist that this relation should hold over each frequency increment, then

$$M_x(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\omega} S_x(u) du = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\omega} |X_T(u)|^2 du. \quad (18)$$

$M_x(\omega)$ is known as the *cumulative power spectrum*. Now, interchange the order of the limiting operator and the integration (assuming it is valid)

$$2\pi M_x(\omega) = \int_{-\infty}^{\omega} S_x(u) du = \int_{-\infty}^{\omega} \lim_{T \rightarrow \infty} \frac{|X_T(u)|^2}{T} du. \quad (19)$$

If $M_x(\omega)$ is differentiable, then

$$2\pi \frac{dM_x(\omega)}{d\omega} = S_x(\omega). \quad (20)$$

Under these conditions

$$S_x(\omega) = \lim_{T \rightarrow \infty} \frac{|X_T(\omega)|^2}{T} \quad (21)$$

Taking the inverse Fourier transform of Equation (21) gives us

$$\mathcal{F}^{-1}\{S_x(\omega)\} = \frac{1}{2\pi} \int_{-\infty}^{\infty} \lim_{T \rightarrow \infty} \frac{|X_T(\omega)|^2}{T} e^{j\omega\tau} d\omega. \quad (22)$$

Interchanging the order of operation yields

$$\begin{aligned}
 \mathcal{F}^{-1}\{S_x(\omega)\} &= \lim_{T \rightarrow \infty} \frac{1}{2\pi T} \int_{-\infty}^{\infty} X_T^*(\omega) X_T(\omega) e^{j\omega\tau} d\omega \\
 &= \lim_{T \rightarrow \infty} \frac{1}{2\pi T} \int_{-\infty}^{\infty} \int_{-T/2}^{T/2} x^*(t) e^{j\omega t} dt \int_{-T/2}^{T/2} x(t') e^{-j\omega t'} dt' e^{j\omega\tau} d\omega \\
 &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x^*(t) \int_{-T/2}^{T/2} x(t') \left[\frac{1}{2\pi} \int_{-\infty}^{\infty} e^{j\omega(t-t'+\tau)} d\omega \right] dt' dt.
 \end{aligned} \tag{23}$$

The integration over ω in the above equation is equal to $\delta(t-t'+\tau)$, therefore

$$\begin{aligned}
 \mathcal{F}^{-1}\{S_x(\omega)\} &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x^*(t) \int_{-T/2}^{T/2} x(t') \delta(t-t'+\tau) dt' dt \\
 &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x^*(t) x(t+\tau) dt
 \end{aligned} \tag{24}$$

The inverse Fourier transform of $S_x(\omega)$ is called *autocorrelation function* of $x(t)$ and is denoted by $R_x(\tau)$.

To summarize

$$R_x(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x^*(t) x(t+\tau) dt \tag{25}$$

and

$$S_x(\omega) = \mathcal{F}\{R_x(\tau)\} \tag{26}$$

If the signal is periodic with period T_0 then,

$$R_x(\tau) = \frac{1}{T_0} \int_{-T_0}^{T_0} x^*(t) x(t+\tau) dt \tag{27}$$

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2 Random Signals and Noise

Basic Definitions

- Define an *experiment* with random *outcome*.
- Mapping of the outcome to a variable \Rightarrow random variable.
- Mapping of the outcome to a function \Rightarrow random function.

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Probability (Cumulative) Distribution Function (cdf)

$$F_X(x) = \text{probability that } X \leq x = P(X \leq x) \tag{28}$$

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Probability Density Function (pdf)

$$f_X(x) = \frac{dF_X(x)}{dx} \tag{29}$$

and

$$P(x_1 < X \leq x_2) = F_X(x_2) - F_X(x_1) = \int_{x_1}^{x_2} f_X(x) dx \tag{30}$$

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2.1 Statistical Averages

Mean of a Discrete RV

$$\bar{X} = \mathcal{E}[X] = \sum_{j=1}^M x_j P_j \quad (31)$$

Mean of a Continuous RV

$$\bar{X} = \mathcal{E}[X] = \int_{-\infty}^{\infty} x f_X(x) dx \quad (32)$$

Variance of a RV

$$\sigma_X^2 \triangleq \mathcal{E} \{ [X - \mathcal{E}(X)]^2 \} = \mathcal{E}[X^2] - \mathcal{E}^2[X] \quad (33)$$

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Given a two random variables X and Y .

Covariance

$$\mu_{XY} = \mathcal{E} \{ [X - \bar{x}][Y - \bar{Y}] \} = \mathcal{E}[XY] - \mathcal{E}[X]\mathcal{E}[Y] \quad (34)$$

Correlation Coefficient

$$\rho_{XY} = \frac{\mu_{XY}}{\sigma_X \sigma_Y} \quad (35)$$

Autocorrelation

$$R_X(\tau) = \mathcal{E}[X(t)X(t + \tau)] \quad (36)$$

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2.2 Stochastic Processes

Terminology

See Figure 1

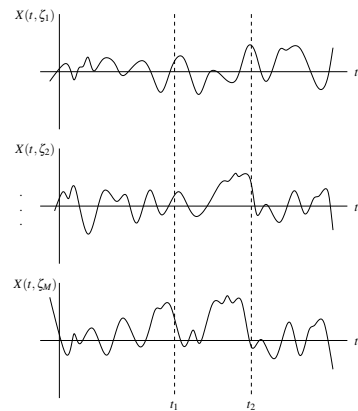


Figure 1: Sample functions of a random process

- $X(t, \zeta_i)$: sample function.
- The governing experiment: random or stochastic process.
- All sample functions: ensemble.
- $X(t_j, \zeta)$: random variable.

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Strict Sense Stationarity

If the joint pdfs depend only on the time difference regardless of the time origin, then the random process is known as *stationary*.

For stationary process means and variances are independent of time and the covariance depends only on the time difference.

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Wide Sense Stationarity

If the joint pdfs depends on the time difference but the mean and variances are time-independent, then the random process is known as *wide-sense-stationary*.

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Ergodicity

If the time statistics equals ensemble statistics, then the random process is known as *ergodic*.

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2.3 Correlation and Power Spectral Density

Power Spectral Density

Given a sample function $X(t, \zeta_i)$ of a random process, we first obtain the power spectral density by means of the Fourier transform of a truncated version $X_T(t, \zeta_i)$ defined as

$$X_T(t, \zeta_i) = \begin{cases} X(t, \zeta_i), & |t| < \frac{1}{2}T \\ 0, & \text{otherwise} \end{cases} \quad (37)$$

The Fourier transform of $X_T(t, \zeta_i)$ is

$$\mathcal{F}\{X_T(t, \zeta_i)\} = \int_{-T/2}^{T/2} X(t, \zeta_i) e^{j2\pi ft} dt \quad (38)$$

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Power Spectral Density of a Random Process

The energy spectral density is $|\mathcal{F}\{X_T(t, \zeta_i)\}|^2$ and the average power density over the T is $|\mathcal{F}\{X_T(t, \zeta_i)\}|^2/T$. Since we have many sample functions, it is intuitive to take the ensemble average as $T \rightarrow \infty$, therefore the power spectral density, $S_X(f)$ is given by

$$S_X(f) = \lim_{T \rightarrow \infty} \frac{|\overline{\mathcal{F}\{X_T(t, \zeta_i)\}}|^2}{T} \quad (39)$$

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Wiener-Khinchine Theorem

$$\begin{aligned} |\mathcal{F}[X_{2T}(t)]|^2 &= \left| \int_{-T}^T X(t) e^{-j\omega t} dt \right|^2 \\ &= \int_{-T}^T \int_{-T}^T X(t) X(\sigma) e^{j\omega(t-\sigma)} dt d\sigma \end{aligned} \quad (40)$$

$$\begin{aligned} \mathcal{E}\{|\mathcal{F}[X_{2T}(t)]|^2\} &= \int_{-T}^T \int_{-T}^T \mathcal{E}[X(t) X(\sigma)] e^{j\omega(t-\sigma)} dt d\sigma \\ &= \int_{-T}^T \int_{-T}^T R(t-\sigma) e^{j\omega(t-\sigma)} dt d\sigma \end{aligned} \quad (41)$$

Apply the change of variables $u = t - \sigma$ and $v = t$, thus (refer to Figure 2).

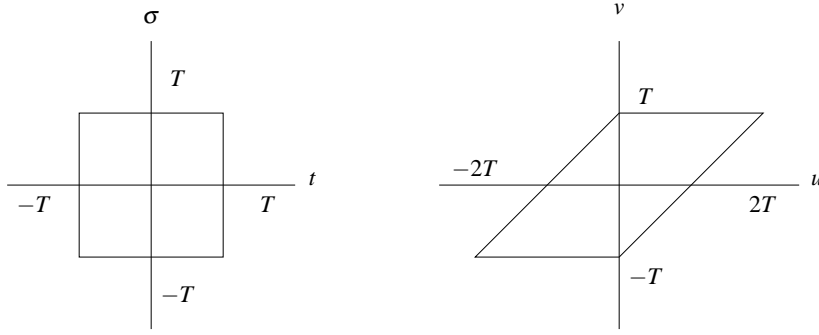


Figure 2: Sample functions of a random process

$$\begin{aligned}
 & \mathcal{E} \{ |\mathcal{F}[X_{2T}(t)]|^2 \} \\
 &= \int_{u=-2T}^0 R(u) e^{-j\omega u} du \left(\int_{-T}^{u+T} dv \right) du + \int_{u=0}^{2T} R(u) e^{-j\omega u} \left(\int_{u-T}^T dv \right) du \\
 &= \int_{-2T}^0 (2T+u) R(u) e^{-j\omega u} du + \int_0^{2T} (2T-u) R(u) e^{-j\omega u} du \\
 &= 2T \int_{-2T}^{2T} \left(1 - \frac{|u|}{2T} \right) R(u) e^{-j\omega u} du
 \end{aligned} \tag{42}$$

$$S_X(f) = \lim_{T \rightarrow \infty} \int_{-2T}^{2T} \left(1 - \frac{|u|}{2T} \right) R_X(u) e^{-j\omega u} du \tag{43}$$

as $T \rightarrow \infty$

$$S(f) \xrightarrow{\mathcal{F}} R(\tau) \tag{44}$$

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2.4 Input-Output Relationship of Linear Systems

$$S_Y(f) = |H(f)|^2 S_X(f) \tag{45}$$

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3 Examples

Example 1 - Mean and Variance

Given a random variable described by the following uniform pdf

$$f_X(x) = \begin{cases} \frac{1}{b-a}, & a \leq x \leq b \\ 0, & \text{otherwise} \end{cases} \tag{46}$$

Compute the mean, the second moment, and the variance.

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Example 2 - Time and Statistical Averages

Given a the following random process

$$X(t) = A \cos(2\pi f_0 t + \Theta) \tag{47}$$

where f_0 is a constant and Θ is a random variable with the following pdf

$$f_\Theta(x) = \begin{cases} \frac{1}{2\pi}, & |\theta| \leq \pi \\ 0, & \text{otherwise.} \end{cases} \tag{48}$$

Compute the statistical and time averages of the first and second moments. Is this process stationary? Is it ergodic?

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Example 3 - Power Spectral Density

Given the same process shown in Example 2, compute the power spectral density using Eq. 39. Verify your answer using Wiener-Khinchine Theorem.

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