

Power Spectral Density and Correlation*

In an analogy to the energy signals, let us define a function that would give us some indication of the relative power contributions at various frequencies, as $S_f(\omega)$. This function has units of power per Hz and its integral yields the power in $f(t)$ and is known as *power spectral density* function. Mathematically,

$$P = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_f(\omega) d\omega. \quad (1)$$

Assume that we are given a signal $f(t)$ and we truncate it over the interval $(-T/2, T/2)$. This truncated version is $f(t)\Pi(t/T)$. If $f(t)$ is finite over the interval $(-T/2, T/2)$, then the truncated function $f(t)\Pi(t/T)$ has finite energy and its Fourier transform $F_T(\omega)$ is

$$F_T(\omega) = \mathcal{F}\{f(t)\Pi(t/T)\}. \quad (2)$$

Parseval's theorem of the truncated version is

$$\int_{-T/2}^{T/2} |f(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |F_T(\omega)|^2 d\omega. \quad (3)$$

Therefore, the average power P across a one-ohm resistor is given by

$$P = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |f(t)|^2 dt = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\infty} |F_T(\omega)|^2 d\omega. \quad (4)$$

Combining Equations (1) and (4), we get

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} S_f(\omega) d\omega = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\infty} |F_T(\omega)|^2 d\omega. \quad (5)$$

In addition if we insist that this relation should hold over each frequency increment, then

$$M_f(\omega) = \frac{1}{2\pi} \int_{-\infty}^{\omega} S_f(u) du = \lim_{T \rightarrow \infty} \frac{1}{T} \frac{1}{2\pi} \int_{-\infty}^{\omega} |F_T(u)|^2 du. \quad (6)$$

$M_f(\omega)$ is known as the *cumulative power spectrum*. Now, interchange the order of the limiting operator and the integration (assuming it is valid)

$$2\pi M_f(\omega) = \int_{-\infty}^{\omega} S_f(u) du = \int_{-\infty}^{\omega} \lim_{T \rightarrow \infty} \frac{|F_T(u)|^2}{T} du. \quad (7)$$

If $M_f(\omega)$ is differentiable, then

$$2\pi \frac{dM_f(\omega)}{d\omega} = S_f(\omega). \quad (8)$$

Under these conditions

$$S_f(\omega) = \lim_{T \rightarrow \infty} \frac{|F_T(\omega)|^2}{T} \quad (9)$$

*This derivation are adapted from F.G. Stremler, *Introduction to Communication Systems*, 2nd Ed., Addison-Wesley, Massachusetts, 1982

Taking the inverse Fourier transform of Equation (9) gives us

$$\mathcal{F}^{-1}\{S_f(\omega)\} = \frac{1}{2\pi} \int_{-\infty}^{\infty} \lim_{T \rightarrow \infty} \frac{|F_T(\omega)|^2}{T} e^{j\omega\tau} d\omega. \quad (10)$$

Interchanging the order of operation yields

$$\begin{aligned} \mathcal{F}^{-1}\{S_f(\omega)\} &= \lim_{T \rightarrow \infty} \frac{1}{2\pi T} \int_{-\infty}^{\infty} F_T^*(\omega) F_T(\omega) e^{j\omega\tau} d\omega \\ &= \lim_{T \rightarrow \infty} \frac{1}{2\pi T} \int_{-\infty}^{\infty} \int_{-T/2}^{T/2} f^*(t) e^{j\omega t} dt \int_{-T/2}^{T/2} f(t') e^{-j\omega t'} dt' e^{j\omega\tau} d\omega \\ &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} f^*(t) \int_{-T/2}^{T/2} f(t') \left[\frac{1}{2\pi} \int_{-\infty}^{\infty} e^{j\omega(t-t'+\tau)} d\omega \right] dt' dt. \end{aligned} \quad (11)$$

The integration over ω in the above equation is equal to $\delta(t - t' + \tau)$, therefore

$$\begin{aligned} \mathcal{F}^{-1}\{S_f(\omega)\} &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} f^*(t) \int_{-T/2}^{T/2} f(t') \delta(t - t' + \tau) dt' dt \\ &= \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} f^*(t) f(t + \tau) dt \end{aligned} \quad (12)$$

The inverse Fourier transform of $S_f(\omega)$ is called *autocorrelation function* of $f(t)$ and is denoted by $R_f(\tau)$.

To summarize

$$R_f(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} f^*(t) f(t + \tau) dt \quad (13)$$

and

$$S_f(\omega) = \mathcal{F}\{R_f(\tau)\} \quad (14)$$

If the signal is periodic with period T_0 then,

$$R_f(\tau) = \frac{1}{T_0} \int_{-T_0}^{T_0} f^*(t) f(t + \tau) dt \quad (15)$$