

3.23 (a)(b) Determine the inverse DTFT of the following DTFTs:

$$\begin{aligned}
 \text{a) } H_1(e^{j\omega}) &= 1 + 2 \cos \omega + 3 \cos 2\omega \\
 &= 1 + 2 \left(\frac{e^{j\omega} + e^{-j\omega}}{2} \right) + 3 \left(\frac{e^{j2\omega} + e^{-j2\omega}}{2} \right) \\
 &= 1 + e^{j\omega} + e^{-j\omega} + \frac{3}{2} e^{j2\omega} + \frac{3}{2} e^{-j2\omega}
 \end{aligned}$$

$$\{h_1[n]\} = \left\{ \frac{3}{2}, 1, \underset{\uparrow}{1}, 1, \frac{3}{2} \right\}$$

$$\begin{aligned}
 \text{b) } H_2(e^{j\omega}) &= (3 + 2 \cos \omega + 4 \cos 2\omega) \cos(\omega/2) e^{-j\omega/2} \\
 &= \left[3 + 2 \left(\frac{e^{j\omega} + e^{-j\omega}}{2} \right) + 4 \left(\frac{e^{j2\omega} + e^{-j2\omega}}{2} \right) \right] \left(\frac{e^{j\omega/2} + e^{-j\omega/2}}{2} \right) e^{-j\omega/2} \\
 &= \frac{1}{2} [1 + e^{-j\omega}] (3 + e^{j\omega} + e^{-j\omega} + 2e^{j2\omega} + 2e^{-j2\omega}) \\
 &= \frac{1}{2} [4 + 3e^{j\omega} + 4e^{-j\omega} + 3e^{-j2\omega} + 2e^{j2\omega} + 2e^{-j3\omega}] \\
 &= 2 + \frac{3}{2} e^{j\omega} + 2e^{-j\omega} + \frac{3}{2} e^{-j2\omega} + e^{j2\omega} + e^{-j3\omega}
 \end{aligned}$$

$$\{h_2[n]\} = \left\{ 1, \frac{3}{2}, \underset{\uparrow}{2}, 2, \frac{3}{2}, 1 \right\}$$

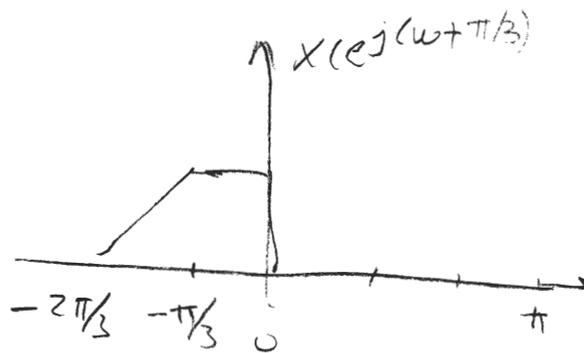
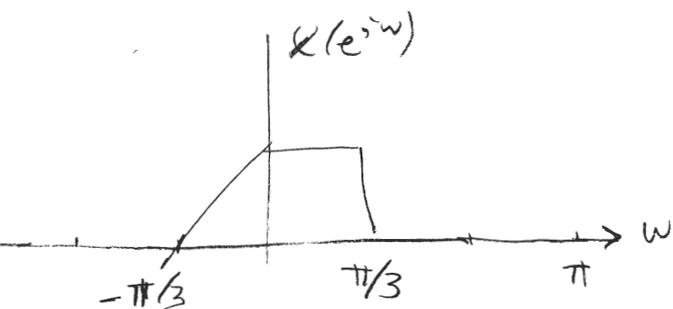
3.31 (a)/(b) Without computing the DTFT, determine which of the following sequences have real-valued DTFT and which imaginary-valued DTFTs:

$$\text{a) } x_1[n] = \begin{cases} n, & -N \leq n \leq N \\ 0, & \text{o.w} \end{cases}$$

Since $x_1[n]$ is real and odd then the DTFT is imaginary.

Since $x_2[n]$ is real and even, then the DTFT is real-valued

3.35 | A sequence $x[n]$ has a zero-phase DTFT $X(e^{j\omega})$ as sketched. Sketch the DTFT of the sequence $e^{-j\pi n/3} x[n]$.



54 | A causal LTI FIR system is characterized by an impulse response $h[n] = a_1 \delta[n] + a_2 \delta[n-1] + a_3 \delta[n-2] + a_4 \delta[n-3] + a_5 \delta[n-4]$. For what values of the pulse response samples will its frequency response $H(e^{j\omega})$ have a linear phase?

$$\begin{aligned}
 H(e^{j\omega}) &= a_1 + a_2 e^{-j\omega} + a_3 e^{-j2\omega} + a_4 e^{-j3\omega} + a_5 e^{-j4\omega} \\
 &= (a_1 e^{j2\omega} + a_5 e^{-j2\omega}) e^{-j2\omega} + (a_2 e^{j\omega} + a_4 e^{-j\omega}) e^{-j2\omega} \\
 &\quad + a_3 e^{-j2\omega}
 \end{aligned}$$

If $a_1 = a_5$ & $a_2 = a_4$ then

$$H(e^{j\omega}) = (2a_1 \cos(2\omega) + 2a_2 \cos(\omega) + a_3) e^{-j2\omega}$$

↑
Linear phase

3.55 An FIR LTI is described by

$$y[n] = a_1 x[n+k] + a_2 x[n+k-1] + a_2 x[n+k-2] + a_1 x[n+k-3]$$

$$H(e^{j\omega}) = a_1 e^{jk\omega} + a_2 e^{j(k-1)\omega} + a_2 e^{j(k-2)\omega} + a_1 e^{j(k-3)\omega}$$

Determine the frequency for its freq response $H(e^{j\omega})$. For what values of k will the system have a freq response that is a real $f(\omega)$

$$= e^{jk\omega} (a_1 + a_2 e^{-j\omega} + a_2 e^{-j2\omega} + a_1 e^{-j3\omega})$$

$$= e^{j(k-3/2)\omega} [a_1 e^{j3/2\omega} + a_1 e^{-j3/2\omega} + a_2 e^{j1/2\omega} + a_2 e^{-j1/2\omega}]$$

$$= e^{j(k-3/2)\omega} [2a_1 \cos(3/2\omega) + 2a_2 \cos(1/2\omega)]$$

$H(e^{j\omega})$ will be real if $k = 3/2$